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# High-Speed CMOS Bus Exchange Switches with Active Termination (Bus Hold)

QS3388

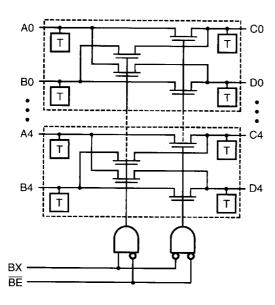
#### **FEATURES/BENEFITS**

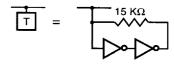
- $5\Omega$  switches connect inputs to outputs
- Active termination drives bus pins to rails when off
- Zero propagation delay
- Undershoot Clamp diodes on all inputs
- Available in 24-pin DIP, SOIC (SO) and QSOP
- Low power CMOS proprietary technology
- Bus exchange allows nibble swap
- Zero ground bounce in flow-through mode
- TTL-compatible input and output levels

#### **DESCRIPTION**

The QS3388 provides two sets of five high-speed CMOS TTL compatible bus switches with active terminators on the bus switch I/O pins. The low ON resistance ( $5\Omega$ ) of the 3388 allows inputs to be connected to outputs without adding propagation delay and without generating additional ground bounce noise. When the switches are turned off, a low drive active terminator circuit drives the disconnected pins to Vcc or ground, away from the TTL threshold. This moves undriven buses from the threshold region to a TTL HIGH or LOW, reducing system noise and power dissipation. The Bus Enable ( $\overline{\text{BE}}$ ) signal turns the switches on. The Bus Exchange (BX) signal provides nibble swap of the AB and CD pairs of signals. This exchange configuration allows byte swapping of buses in systems. It can also be used as a five 2-to-1 multiplexer and to create low delay barrel shifters, etc.

#### **FUNCTIONAL BLOCK DIAGRAM**





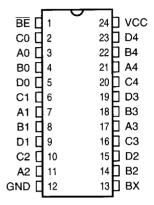
MDSL-00033-02

#### **PIN DESCRIPTION**

Name	1/0	Function	
A4-A0, B4-B0	1/0	Buses A, B	
C4-C0, D4-D0	I/O	Buses C, D	
BE	ı	Bus Switch Enable	
вх	ı	Bus Exchange	

# PIN CONFIGURATION (All Pins Top View)

PDIP, SOIC (SO), QSOP



### **FUNCTION TABLE**

BE	вх	A4-A0	B4-B0	Function
Н	Х	Hi-Z	Hi-Z	Disconnect
L	L	C4-C0	D4-D0	Connect
L	Н	D4-D0	C4-C0	Exchange

#### **ABSOLUTE MAXIMUM RATINGS**

Supply Voltage to Ground	0.5V to +7.0V
DC Switch Voltage Vs	
DC Input Voltage Vin	
AC Input Voltage (for a pulse width ≤ 20 ns)	
DC Output Current Max. Sink Current/Pin	120 mA
Maximum Power Dissipation	0.5 watts
Tstg Storage Temperature	–65° to +150°C

Note: Stresses greater than those listed under ABSOLUTE MAXIMUM RATINGS may cause permanent damage to QSI devices that result in functional or reliability type failures.

#### **CAPACITANCE**

 $T_A = 25^{\circ}C$ , f = 1 MHz,  $V_{IN} = 0V$ ,  $V_{OUT} = 0V$ 

	S	SOIC		QSOP		IP	
Pins	Тур	Max	Тур	Max	Тур	Max	Unit
Control Pins	3	4	3	4	4	5	pF
QuickSwitch Channels	7	8	7	8	8	9	рF

Note: Capacitance is characterized but not tested.

## DC ELECTRICAL CHARACTERISTICS OVER OPERATING RANGE

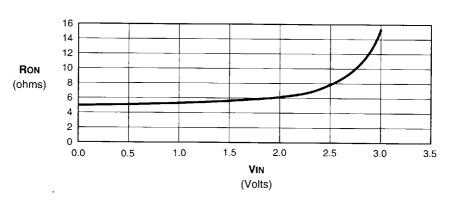
Commercial: Ta =  $0^{\circ}$ C to  $70^{\circ}$ C,  $V_{CC} = 5.0V \pm 5\%$ 

Symbol	Parameter	Test Conditions	Min	Typ <sup>(1)</sup>	Max	Unit
Vıн	Input HIGH Voltage	Guaranteed Logic HIGH for Control Inputs	2.0	_		٧
VIL	Input LOW Voltage	Guaranteed Logic LOW for Control Inputs	-	_	0.8	٧
l lin l	Input Leakage Current(2)	0 ≤ Vin ≤ Vcc		_	5	μА
lτ	Input Current, Disconnect	Vout = 0.5V, 4.5V; Vcc = 5V	15	33	70	μА
Rт	Terminator Resistance(6,7)		_	15		ΚΩ
los	Short Circuit Current(3)	AB(CD) = 0V, CD(AB) = Vcc	_	300		mA
Ron	Switch ON Resistance <sup>(4,5)</sup>	Vcc = Min., Vin = 0.0V Ion = 30 mA	_	5	7	Ω
Ron	Switch ON Resistance <sup>(4,5)</sup>	Vcc = Min., ViN = 2.4V IoN = 15 mA	_	10	15	Ω

#### Notes:

- 1. Typical values indicate Vcc = 5.0V and TA = 25°C.
- 2. During input/output leakage, testing all pins are at a HIGH or LOW state, and the  $\overline{\text{BE}}$  control is HIGH.
- Not more than one output should be used to test this high power condition and the duration is ≤1 second.
- Measured by voltage drop between AB and CD pin at indicated current through the switch. ON resistance is determined by the lower of the voltages on the two (A or B, C or D) pins.
- 5. Max. value Ron guaranteed but not tested.
- 6. Characterized but not tested.
- 7. Computed from parameter "IT" and the test conditions.

# Typical ON Resistance vs V<sub>IN</sub> at 4.75 Vcc



#### **POWER SUPPLY CHARACTERISTICS**

Symbol	Parameter	Test Conditions(1)	Max	Unit
loca	Quiescent Power Supply Current	Vcc = Max., Vin = GND or Vcc, f = 0	1.5	μА
ΔΙσο	Power Supply Current per Input HIGH(2)	Vcc = Max., Vin = 3.4V, f = 0 per Control Input	2.5	mA
Qccd	Dynamic Power Supply Current per MHz <sup>(3)</sup>	Vcc = Max., A, B, C, D Pins Open, Control Inputs Toggling @ 50% Duty Cycle	0.25	mA/ MHz
lc	Total Power Supply Current <sup>(4,5)</sup>	Vcc = Max., A, B, C, D Pins at 0.0V, Control Inputs Toggling @ 50% Duty Cycle Vih = 3.4V, f Clock + MHz	9.0	mA

#### Notes:

- 1. For conditions shown as Min. or Max., use the appropriate values specified under DC specifications.
- 2. Per TTL driven input (VIN = 3.4V, control inputs only). A, B, C, D pins do not contribute to lcc.
- 3. This current applies to the control inputs only and represents the current required to switch internal capacitance at the specified frequency. The A, B, C, D inputs generate no significant AC or DC currents as they transition. This parameter is guaranteed by design, but not tested.
- 4. Ic = I Quiescent + I Inputs + I Dynamic.
  - $Ic = Icc + \Delta IccDhNT + Qccd (fiNi).$
  - Icc = Power Supply Current for each TTL HIGH input (VIN = 3.4V, control inputs only).
  - DH = Duty Cycle for each TTL input that is HIGH (control inputs only).
  - NT = Number of TTL inputs that are at DH (control inputs only).
  - fi = frequency that the inputs are toggled (control inputs only).
- 5. Note that activity on A, B, C, D inputs do not contribute to Ic if A, B, C, D inputs are between GND and Vcc. The switches merely connect and pass through activity on these pins. For example: If the control inputs are at 0V and the switches are on, Ic will be equal to Icc only regardless of activity on the A, B, C, D pins.

## SWITCHING CHARACTERISTICS OVER OPERATING RANGE

Commercial: Ta = 0°C to 70°C, Vcc = 5.0V  $\pm$  5% CLOAD = 50 pF, RLOAD = 500 $\Omega$  unless otherwise noted.

Symbol	Description <sup>(1)</sup>	Min	Тур	Max	Unit
tplH tpHL	Data Propagation Delay <sup>(2,3)</sup> AiBi to CiDi, CiDi to AiBi	_		0.25(3)	ns
tpzL tpzH	Switch Turn-on Delay <sup>(1)</sup> BE to Ai, Bi, Ci, Di	1.5		6.5	ns
tplz tphz	Switch Turn-off Delay <sup>(1,2)</sup> BE to Ai, Bi, Ci, Di	1.5		5.5	ns
tвх	Switch Multiplex Delay <sup>(1)</sup> BX to Ai, Bi, Ci, Di <sup>(1)</sup>	1.5	_	6.5	ns
I Qci I	Charge Injection <sup>(2,4)</sup>	_	1.5	_	pC
I Qoci I	Differential Charge Injection <sup>(2,5)</sup>		<0.5	_	pC

#### Notes:

- 1. See Test Circuit and Waveforms. Minimums guaranteed but not tested.
- 2. This parameter is guaranteed by design but not tested.
- 3. The bus switch contributes no propagation delay other than the RC delay of the ON resistance of the switch and the load capacitance. The time constant for the switch alone is of the order of 0.25 ns for 50 pF. Since this time constant is much smaller than the rise/fall times of typical driving signals, it adds very little propagation delay to the system. Propagation delay of the bus switch when used in a system is determined by the driving circuit on the driving side of the switch and its interaction with the load on the driven side.
- 4. Measured at switch turn off, A to C, load = 50 pF in parallel with 10 meg scope probe, VIN at A = 0.0V.
- 5. Measured at switch turn off through bus multiplex, A to C ≥ A to D, B connected to C, load = 50 pF in parallel with 10 meg scope probe, V<sub>IN</sub> at A = 0.0V. Charge injection is reduced because the injection from the turn off of the A to C switch is compensated by the turn on of the B to C switch.

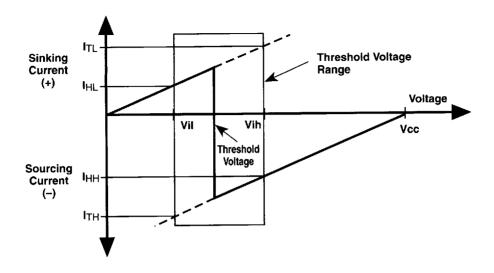


Figure 1: Trip and Hold Current Characteristics of the Last Value Latch

TTL Logic device datasheets specify V<sub>IL</sub> and V<sub>IH</sub> values which guarantee that HIGH and LOW logic level inputs will be interpreted correctly. V<sub>IL</sub> and V<sub>IH</sub> specifications accommodate variations over temperature and voltage of the inherent switching point of the device, commonly referred to as the threshold voltage. Threshold voltage could theoretically exist anywhere between V<sub>IL</sub> and V<sub>IH</sub>.

Figure 1 shows V-I characteristics of the terminator circuit. It demonstrates that as a forced input voltage increases from 0V (logic LOW), the sink current increases linearly. Conversely, if a current is forced into the device, then terminator voltage will gradually increase until VIL is violated. This current value is the hold current for logic LOW (IHL). Higher values of sink current will violate system VIL levels. At some point, if sink current increases further, the terminator voltage will reach the switching threshold and the device will switch to logic HIGH. Since the maximum possible value of the threshold (allowed by spec) is VIH, maximum trip current (ITL) is specified as the current required to force the device up to VIH.

If a terminator input signal is at  $V_{CC}$ , the node current is 0  $\mu$ A, but if  $V_{IN}$  is forced downward, the node current gradually increases. If enough current is pulled from the terminator, system  $V_{IH}$  will be violated (IHH). Since the minimum possible switching threshold is  $V_{IL}$ , ITH is defined as the current required for the high signal to be forced to  $V_{IL}$ . Note that ITL and ITH are theoretical numbers. They will never be reached in practical situations since threshold voltage always lies between the  $V_{IL}$  and  $V_{IH}$  specifications.